(1) z-parameter "open ct. parameter"

$V_1 = Z_{11} I_1 + Z_{12} I_2$

$V_2 = Z_{21} I_1 + Z_{22} I_2$

$$\begin{pmatrix} V_1 \\ V_2 \end{pmatrix} = \begin{pmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \end{pmatrix}$$

Z-Matrix

---

(2) Y-Parameter "short ct. parameter"

$I_1 = Y_{11} V_1 + Y_{12} V_2$

$I_2 = Y_{21} V_1 + Y_{22} V_2$

$$\begin{pmatrix} I_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \end{pmatrix}$$

Y-Matrix

---

(3) Hybrid parameter

$V_1 = h_{11} I_1 + h_{12} V_2$

$I_2 = h_{21} I_1 + h_{22} V_2$

$$\begin{pmatrix} V_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{pmatrix} \begin{pmatrix} I_1 \\ V_2 \end{pmatrix}$$

---

(4) Transmission parameter.

$V_1 = A V_2 - B I_2$

$I_1 = C V_2 - D I_2$

$$\begin{pmatrix} V_1 \\ I_1 \end{pmatrix} = \begin{pmatrix} A & B \\ C & D \end{pmatrix} \begin{pmatrix} V_2 \\ I_2 \end{pmatrix}$$

* if $Z_{12} = Z_{21}$ or $Y_{12} = Y_{21}$
∴ Reciprocal system

* if $Z_{11} = Z_{22}$ or $Y_{11} = Y_{22}$
∴ symmetrical system

* Z-parameter → T. Model
Y-parameter → π-Model

T-Model & π-Model

. from (T → π)

$$Z_A = \frac{Z_1 Z_2 + Z_1 Z_3 + Z_2 Z_3}{Z_2}$$

$$Z_B = \frac{Z_1 Z_2 + Z_1 Z_3 + Z_2 Z_3}{Z_1}$$

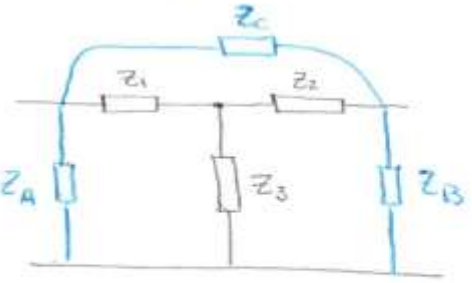
$$Z_C = \frac{Z_1 Z_2 + Z_1 Z_3 + Z_2 Z_3}{Z_3}$$

. from (π → T)

$$Z_1 = \frac{Z_A Z_C}{Z_A + Z_B + Z_C}$$

$$Z_2 = \frac{Z_B Z_C}{Z_A + Z_B + Z_C}$$

$$Z_3 = \frac{Z_A Z_B}{Z_A + Z_B + Z_C}$$



→ Relationship among two port parameter.

[1] $Z \rightarrow Y$

$$[Y] = [Z]^{-1} = \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix}^{-1} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix}$$

---

[2] $Y \rightarrow Z$

$$[Z] = [Y]^{-1}$$

---

[3] $T \rightarrow Z$

$$Z_{11} = \frac{A}{C}$$

$$Z_{12} = \frac{AD}{C} - B$$

$$Z_{21} = \frac{1}{C}$$

$$Z_{22} = \frac{D}{C}$$

---

[3] $Z \rightarrow T$

$$C = \frac{1}{Z_{21}}$$

$$D = \frac{Z_{22}}{Z_{21}}$$

$$A = \frac{Z_{11}}{Z_{21}}$$

$$B = \frac{Z_{11} Z_{22}}{Z_{21}} - Z_{12}$$

---

* check for Hurwitz.

. If $P(s) = s^3 + 2s^2 + 3s + 6$ "Polynomial"

Sol:

even part $m(s) = 2s^2 + 6$

odd part $n(s) = s^3 + 3s$

أقسم ال order الكبير على ال order الصغير
والناتج يطلع ال poles
عدد الـ poles من degree

Ex: $s^7 + 2s^6 + 2s^5 + s^4 + 4s^3 + 8s^2 + 8s + 4$

Sol:- $m(s) = 2s^6 + s^4 + 8s^2 + 4$

$n(s) = s^7 + 2s^5 + 4s^3 + 8s$

$2s^6 + s^4 + 8s^2 + 4 \;\big|\; s^7 + 2s^5 + 4s^3 + 8s$

$q_1 \leftarrow \frac{1}{2}s$: $s^7 + \frac{1}{2}s^5 + 4s^3 + 2s$

$\frac{3}{2}s^5 + 6s \;\big|\; 2s^6 + s^4 + 8s^2 + 4$

$q_2 \leftarrow \frac{4}{3}s$: $2s^6 + 8s^2$

$s^4 + 4 \;\big|\; \frac{3}{2}s^5 + 6s$

$q_3 \leftarrow \frac{3}{2}s$: $\frac{3}{2}s^5 + 6s$

0

Note:- check Herwitz لو طلب

$P(s) = \dfrac{s^2 + 2s + 1}{s^3 + 2s^2 + 2s + 40}$

Sol:-

$\therefore P(s) = N(s) + D(s) = s^3 + 3s^2 + 4s + 41$

not Herwitz مش بنيجي اختبار

- if all quotients are +ve ∴ Herwitz
- if one quotient is -ve ∴ non Herwitz

All are +ve

Ther is only 3 quotient only

∴ There is common factor

$s^3 + 2s^2 + 2s + 1$

$s^4 + 4 \;\big|\; s^7 + 2s^6 + 2s^5 + s^4 + 4s^3 + 8s^2 + 8s + 4$

$s^7 \qquad + 4s^3$

$2s^6 + 2s^5 + s^4 + 8s^2 + 8s + 4$

$2s^6 \qquad + 8s^2$

$2s^5 + s^4 + 8s + 4$

$2s^5 \qquad + 8s$

$s^4 + 4$

$s^4 + 4$

0

بقسم على اكبر حاجة فيهم على فوق
واقسم قسمة مقولة عادي
ظاهرها زي بتاعة فوق.

$(s^4 + 4)(s^3 + 2s^2 + 2s + 1) = 0$

$s^4 = -4$

$s^3 = \pm j2$ ← هيطلع ليا تلات زوج

$s = \pm\sqrt{2}j \quad \pm\sqrt{2}j$

∴ There is a multi poles at $j\omega$ axis
∴ not Herwitz.

لما لو كان طلع بس مثلا $s = \pm j0$ و القوس
LHS يعني -ve طلع كل الـ Poles اللي طلع
Herwitz شيفي -s.

Test the following fun. wether they are PRF or not giving reasons.

given :- $F(s) = \frac{N(s)}{D(s)}$

∴ condition to be PRF

$m_1 m_2 - n_1 n_2 \geq 0 \quad \forall w$

- all poles lie in LHP
- All coeff are real & +ve

m_1 & m_2 : even part in $N(s)$ & $D(s)$

n_1 & n_2 : odd part in $N(s)$ & $D(s)$.

Note ملحوظة

قبل ما أعمل حاجة لازم أتأكد أنه مفيش حد سبني معرفتين في بعض وزي بعض فوقه أو تحت بعض منلاً لو أدالي

$F(s) = \frac{s(s+5)}{s^4+2s^2+1}$

البسط ده ممكن يبقى

$(s^2+1)(s^2+1)$ بعض متكرر

∴ not PRF because has a multiples

$(s^2+1)(s^2+1) = (s^2+1)^2$

ex: $F(s) = \frac{2s^3+3s^2+3s+3}{s^2+1}$

Sol: Hurwitz هلازم أول حاجة check لازم أول

$m_1 = 3s^2+3 \qquad n_1 = 2s^3+3s$

$m_2 = s^2+1 \qquad n_2 = 0$

$m_1 m_2 - n_1 n_2 = (3s^2+3)(s^2+1) - 0$

$= 3s^4+6s^2+3 \Big|_{jw}$

$= 3(jw)^4 - 6(jw)^2 + 3$

$= 3w^4 + 6w + 3 \geq 0 \quad \forall w$ بعوض فرضية Zero ومرة أن رقم

∴ $f(s)$ is PRF :

check PRF قبل ما أعمل check for Hurwitz لازم أعمل

$F'(s) = 2s^3+3s^2+3s+3 + s^2+1$

$= 3s^3 + 4s^2 + 3s + 4$

$m'(s) = 4s^2 + 4$

$n'(s) = 3s^3 + 3s$

واعمل القسمة على الصغير

$4s^2+4 \;\big|\; 3s^3+3s$

$\frac{3}{4}s \quad 3s^3 + 3s$

0

$\frac{3}{4}s + 1$

$4s^2+4 \;\big|\; 3s^3+4s^2+3s+4$

$3s^2 + 3s$

$4s^2 + 4$

$4s^2 + 1$

0

∴ $(4s^2+4)(\frac{3}{4}s+1) = 0$

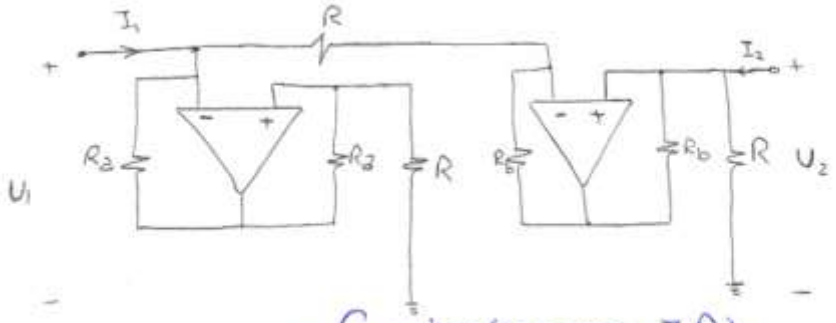
$s = \pm j \qquad s = -4/3$

LHS

∴ Herwit.

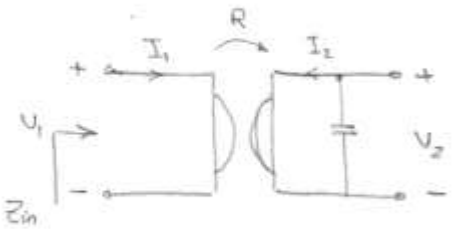
Network Realization

Deboo



≡ Gyrator (2opamp + 7R)

Grounded inductor

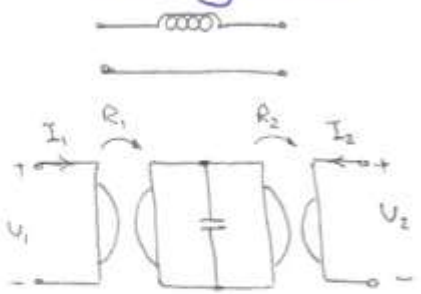


$L = R^2C$

10 element

(2opamp + 7R + 1C)

Floating inductor

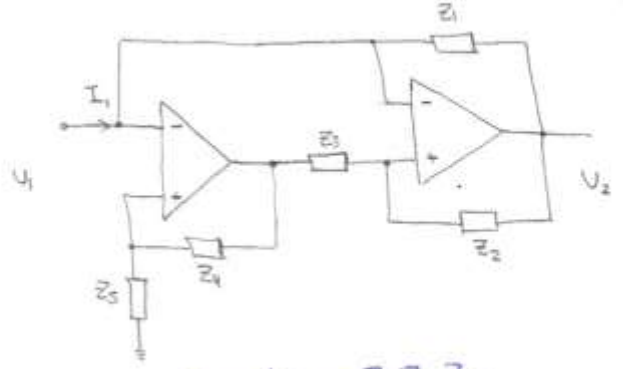


$L = R^2C$

14 element

(4opamp + 14R + 1C)

Riorden



$$Z_{in} = \frac{V_1}{I_1} = \frac{Z_1 Z_3 Z_5}{Z_2 Z_4}$$

FDNR

(2 op-amp + 3R + 2C)

$Z_1 = Z_3 = Z_5 = R$

$Z_2 = Z_4 = 1/sC$

$\therefore Z_{in} = -\omega^2 C^2 R^3$

$Z_2 = Z_4 = Z_5 = R$

$Z_1 = Z_3 = 1/sC$

$\therefore Z_{in} = 1/\omega^2 C^2 R$

Grounded inductor

(2 opamp + 3R + 2C)

$Z_1 = Z_2 = Z_3 = Z_5 = R$

$Z_4 = 1/sC$

$\therefore Z_{in} = R^2 sC$

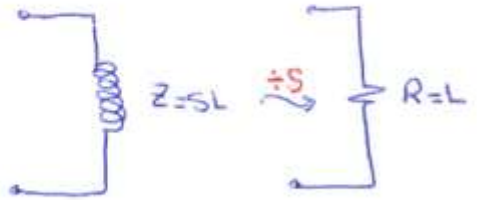
$\therefore L = R^2 C$

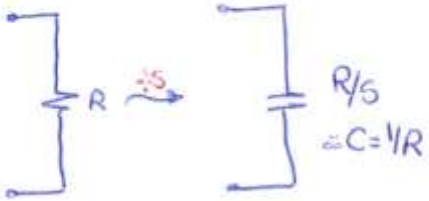
Floating inductor

(2opamp + 7R + 2C)

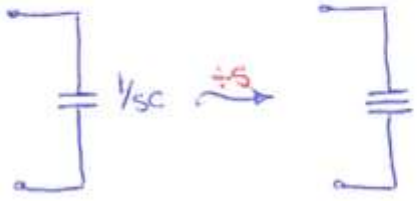
Sheet 4

Bruton transformation





$\therefore C = 1/R$



FDNR

$Z_c = 1/s^2C$

$Z_{in} = 1/\omega^2 C$

Realization of one port Network using LC

Foster I | Foster II | Cauer I | Cauer II

→ Foster (I) " It deals with Z(s)

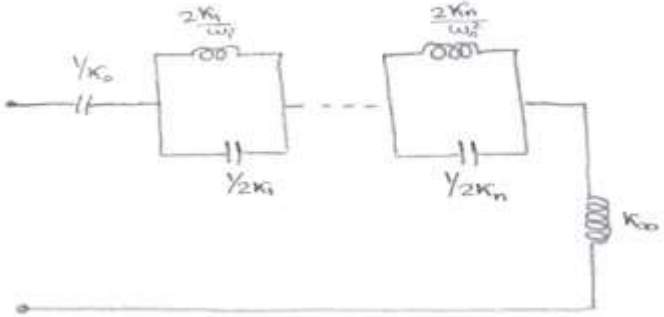
- |Degree of N(s) − Degree of D(s)| ≥ 1
- It provide series Network
- $Z(s) = \frac{odd}{even}$ or $\frac{even}{odd}$

$$\therefore Z(s) = \frac{K_0}{s} + \sum_{n=1}^{a} \frac{2K_r s}{s^2+\omega_r^2} + K_\infty s$$

$$K_0 = \lim_{s\to 0} sZ(s)$$

$$K_\infty = \lim_{s\to \infty} \frac{Z(s)}{s}$$

$$2K_r = \lim_{s^2\to -\omega_r^2} (s^2+\omega_r^2)\frac{Z(s)}{s}$$



→ Foster (II) It deals with y(s)

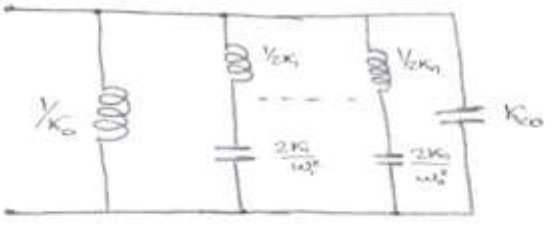
- |Degree of N(s) − Degree of D(s)| ≥ 1
- It provide series Network
- $y(s) = \frac{even}{odd}$ or $\frac{odd}{even}$

$$\therefore y(s) = \frac{K_0}{s} + \sum_{n=1}^{a} \frac{2K_n s}{s^2+\omega_n^2} + K_\infty s$$

$$K_0 = \lim_{s\to 0} sy(s)$$

$$K_\infty = \lim_{s\to \infty} \frac{y(s)}{s}$$

$$2K_r = \lim_{s\to -\omega_r^2} (s^2+\omega_r^2)\frac{y(s)}{s}$$



Cauer (I)

1- It deals with both $Z(s)$ & $Y(s)$

IF Given $Z(s)$

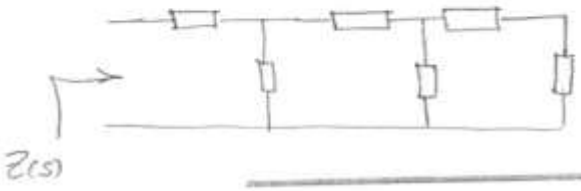
1- $\lim_{s \to \infty} Z(s) = \infty$

If given $Y(s)$

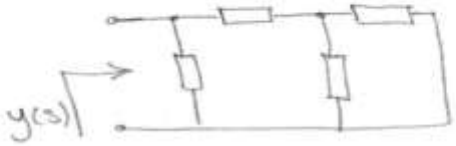
1- $\lim_{s \to 0} Y(s) = \infty$

2- write both $N(s)$ and $D(s)$ in descending order ترتيب هذه
3- Apply continued fraction expansion قسمة بطويلة

4- Draw the resulting network in Ladder form starting with Series element.



5- Draw the resulting network in ladder form start with parallel element.



Cauer(II)

1- It deals with both $Z(s)$ & $Y(s)$

$Y(s)$

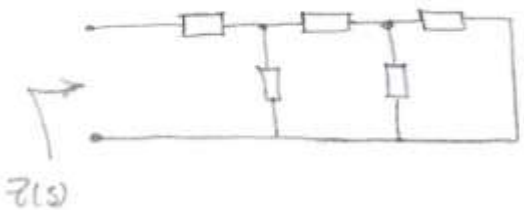
$\lim_{s \to 0} Y(s) = \infty$

1- $Z(s)$

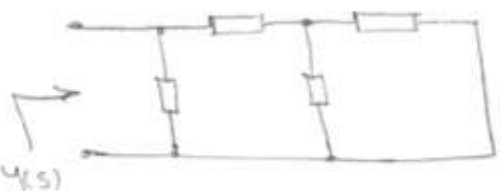
$\lim_{s \to 0} Z(s) = \infty$

2- write both $N(s)$ and $D(s)$ in ascending order
3- Apply continued fraction expansion

4- Draw the resulting Network in ladder form start with Series element



4- Draw the resulting network in ladder form start with parallel element.



s

Realization of one port Network using RC and RL

RC : $Z(s) = \frac{N(s)}{D(s)}$ or $y(s) = \frac{N(s)}{D(s)}$

$\frac{N(s)}{D(s)} = \frac{\text{odd} + \text{even}}{\text{odd} + \text{even}}$ < RC / RL

$Z(0) \geq Z(\infty)$
$y(0) \leq y(\infty)$ } → RC

RL :

$Z(s) = \frac{N(s)}{D(s)}$ or $y(s) = \frac{N(s)}{D(s)}$

$\frac{N(s)}{D(s)} = \frac{\text{odd} + \text{even}}{\text{odd} + \text{even}}$

$Z(0) \leq Z(\infty)$
$y(0) \geq y(\infty)$ } RL

cover I & II

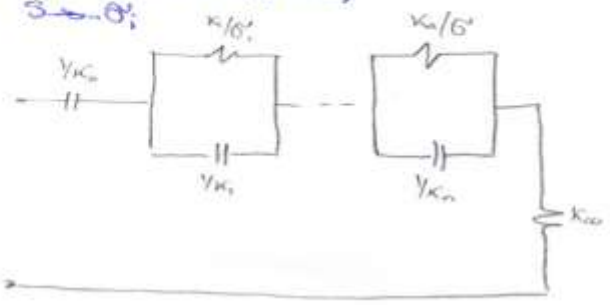
Foster (I)

$$Z(s) = \frac{K_0}{s} + \sum_{i=1} \frac{K_i}{s+\sigma_i} + K_\infty$$

$K_0 = \lim_{s \to 0} sZ(s)$

$K_\infty = \lim_{s \to \infty} Z(s)$

$K_i = \lim_{s \to -\sigma_i} (s+\sigma_i) Z(s)$



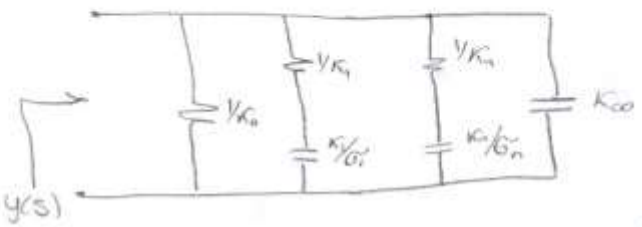
Foster (II)

$$y(s) = \frac{K_0}{s} + \sum_{i=1} \frac{K_i}{s+\sigma_i} + K_\infty$$

$K_0 = \lim_{s \to 0} sy(s)$

$K_\infty = \lim_{s \to \infty} y(s)$

$K_i = \lim_{s \to -\sigma_i} (s+\sigma_i) y(s)$



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